

**22W FRONT REAR OR BRIDGE FULLY PROTECTED
CAR RADIO AMPLIFIER**

SGS-THOMSON

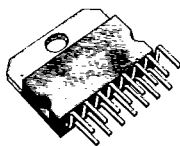
30E D

- HIGH OUTPUT POWER
- POP FREE SWITCHING
- SHORT CIRCUIT PROTECTIONS: R_L SHORT - OUT TO GROUND - OUT TO V_S
- MUTING μP COMPATIBLE
- VERY LOW CONSUMPTION STANDBY
- PROGRAMMABLE TURN ON DELAY
- LOW DISTORTION AND LOW NOISE
- DIFFERENTIAL INPUT

Other Protections :

- LOAD DUMP VOLTAGE SURGE
- LOUDSPEAKER DC CURRENT
- VERY INDUCTIVE LOAD
- OVERRATING TEMPERATURE
- OPEN GROUND

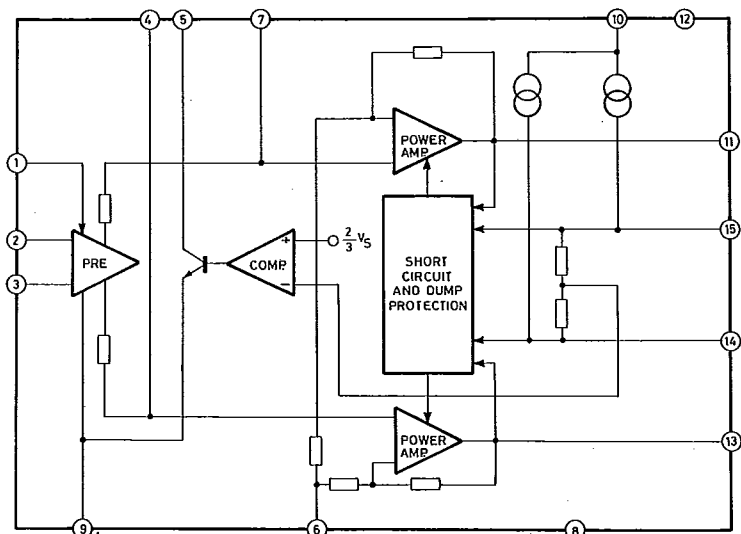
The TDA7255 a class B dual fully protected power amplifier designed for car radio applications. The device can be switched from Front-Rear to Bridge configuration by changing only the loudspeaker connection. An input fader for Front-Rear control is available. A high current capability allows to drive low impedance loads (up to 1.6Ω).



Multiwatt-15

ORDER CODE : TDA7255

BLOCK DIAGRAM



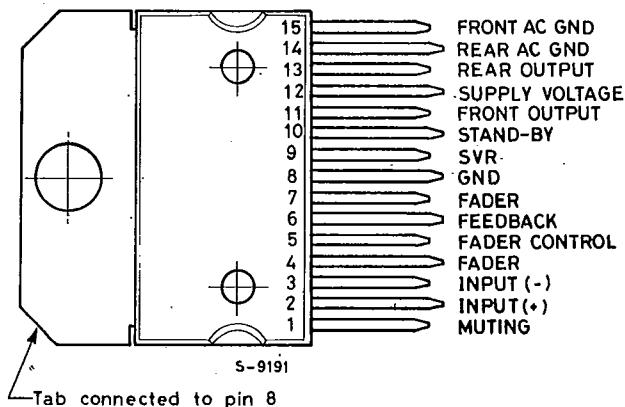
5-9189

ABSOLUTE MAXIMUM RATINGS

| | | | |
|----------------|---|------------|------------|
| V_s | Operating supply voltage | 18 | V |
| $V_{s,DC}$ | DC supply voltage | 28 | V |
| $V_{s,PK}$ | Peak supply voltage (for 50ms) | 40 | V |
| I_o | Output peak current (non repetitive $t = 0.1ms$) | 4.5 | A |
| I_o | Output peak current (repetitive $f \geq 10Hz$) | 4 | A |
| P_{tot} | Power dissipation at $T_{case} = 60^\circ C$ | 30 | W |
| T_{stg}, T_j | Storage and junction temperature | -40 to 150 | $^\circ C$ |

CONNECTION DIAGRAM

(Top view)



THERMAL DATA

| | | | | |
|-----------------|-------------------------------------|-----|----|--------------|
| $R_{th J-case}$ | Thermal resistance junction-case | max | 3 | $^\circ C/W$ |
| $R_{th J-amb}$ | Thermal resistance junction-ambient | max | 40 | $^\circ C/W$ |

SGS-THOMSON

30E D

ELECTRICAL CHARACTERISTICS ($V_s = 14.4V$, $R_L = 4\Omega$, $f = 1KHz$, $T_{amb} = 25^\circ C$ unless otherwise specified)

| Parameter | Test Conditions | Min. | Typ. | Max. | Unit |
|--|-----------------|------|------|------|------------|
| V_s Supply voltage | | 8 | | 18 | V |
| I_d Total quiescent drain current | | | 80 | | mA |
| R_I Input resistance | | | 70 | | K Ω |
| V_i Input saturation voltage | | 300 | | | mV |
| T_j Thermal shut down junction temperature | | | 145 | | $^\circ C$ |

FRONT REAR APPLICATIONS (Fig. 2)

| | | | | | |
|------------------------------|---|-----|-------------------|-----|--------------------|
| P_o Output power | THD = 10% $R_L = 4\Omega$ $R_L = 2\Omega$ $R_L = 1.6\Omega$ | 5.5 | 6.5 11 12.5 | | W W W |
| d Distortion | $P_o = 0.1W$ to 4W | | 0.05 | 0.5 | % |
| G_v Voltage gain | | | 28 | | dB |
| e_N Input noise voltage | $R_G = 10K\Omega$ | | 2.5(**) 2 (*) | | μV μV |
| SVR Supply voltage rejection | $R_G = 100K\Omega$ $V_r = 1V$ $f = 300Hz$ | 36 | 45 | | dB |
| CMR Common mode rejection | | | 55 | | dB |
| η Efficiency | $P_o = 6.5W + 6.5W$ | | 70 | | % |

BRIDGE APPLICATION (Fig. 1)

| | | | | | |
|---------------------------------|--|----|--------------------|-----|--------------------|
| V_{os} Output offset voltage | | | | 250 | mV |
| P_o Output power | THD = 10% $R_L = 4\Omega$ $R_L = 3.2\Omega$ | 18 | 22 25 | | W W |
| d Distortion | $P_o = 0.1W$ to 2W | | 0.05 | | % |
| G_v Voltage gain (C_L) | | | 36 | | dB |
| e_N Total input noise voltage | $R_G = 10K\Omega$ | | 2.5(**) 2.0 (*) | 10 | μV μV |
| η Efficiency | $P_o = 20W$ | | 66 | | % |
| SVR Supply voltage rejection | $R_G = 10K\Omega$, $V_r = 1V$, $f = 300Hz$ | 45 | 58 | | dB |

MUTING AND STAND-BY FUNCTIONS

| | | | | | |
|----------------------------------|-------------------------------------|-----|--|-----|---------|
| Muting attenuation | $V_{ref} = 1W$ $f = 100Hz$ to 10KHz | 60 | | | dB |
| Muting-on threshold voltage | Pin. 1 | 2.4 | | | V |
| Muting-off threshold voltage | Pin. 1 | | | 0.8 | V |
| Stand-by attenuation | $V_{ref} = 1V$ $f = 100Hz$ to 10KHz | 60 | | | dB |
| Stand-by quiescent drain current | | | | 100 | μA |

(**) B = 22Hz to 22KHz

(*) B = curve A

Fig. 1 - Test and application circuit (Bridge amplifier)

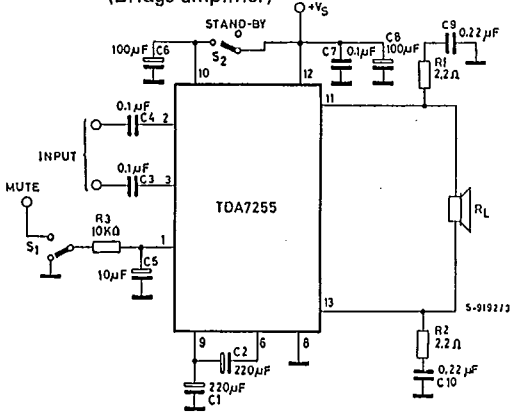
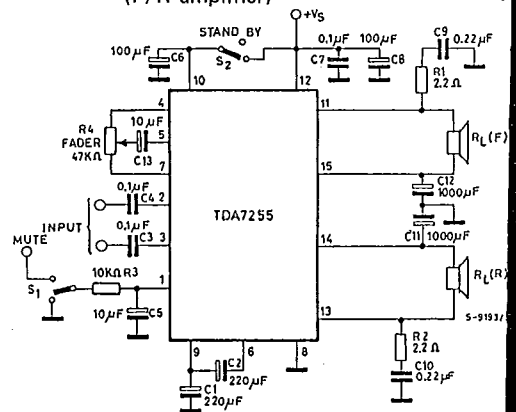
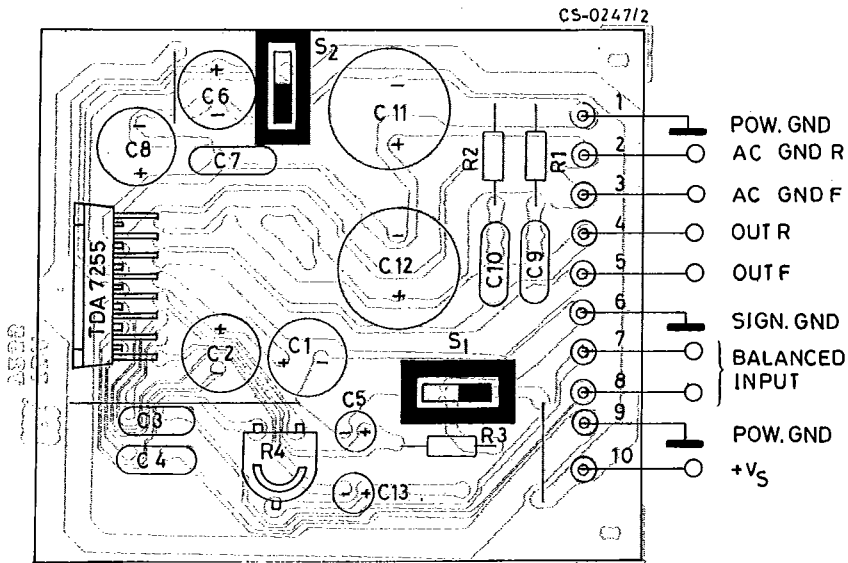


Fig. 2 - Test and application circuit (F/R amplifier)



- Two high impedance inputs available for balanced or unbalanced operation.
- The fader function is automatically inserted in front/rear configuration and allows the distribution of the power between the front and the rear. An external potentiometer must be connected between pins 4 and 7 with the control terminal connected to pin 5 through a decoupling capacitor. In bridge applications the pins 4-5-7 must be left open.
- Turn on delay. The output stages are muted during the turn on transient and start rising after the charge of the capacitor connected between pin 9 and ground. The capacitor also avoids pops during bridge F/R switching.

Fig. 3 - P.C. board and component layout of the circuits of Fig. 1 and 2 (1 : 1 scale)



S G S-THOMSON

30E D

FRONT/REAR CHARACTERISTICS

Fig. 4 - Quiescent drain current vs. supply voltage

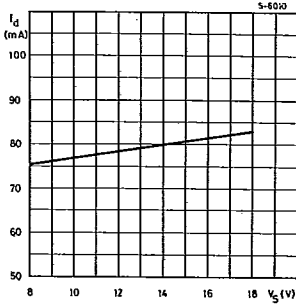


Fig. 5 - Quiescent output voltage vs. supply voltage

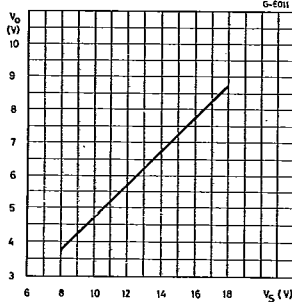


Fig. 6 - Output power vs. supply voltage

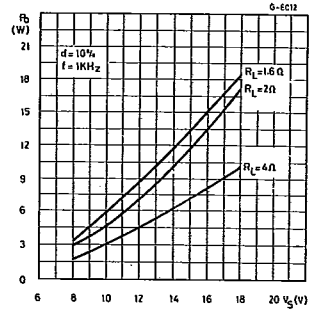


Fig. 7 - Distortion vs. frequency

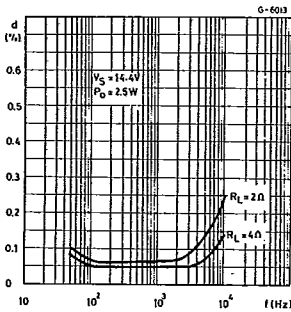


Fig. 8 - Supply voltage rejection vs. capacitor values (C2)

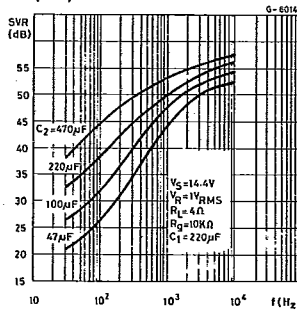


Fig. 9 - Supply voltage rejection vs. capacitor values (C1)

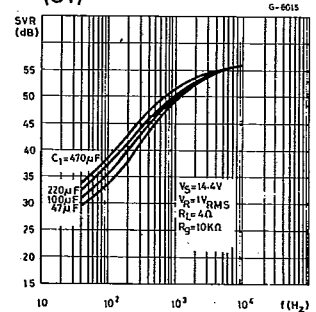


Fig. 10 - Output signal vs. fader control position

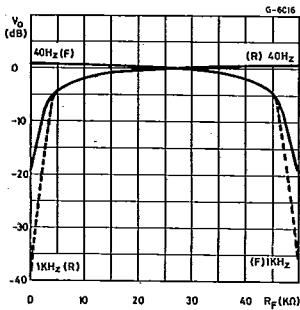


Fig. 11 - Power dissipation and efficiency vs. output power

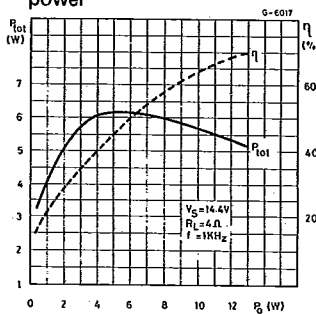
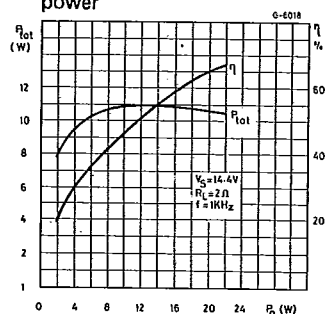


Fig. 12 - Power dissipation and efficiency vs. output power



S G S-THOMSON
BRIDGE CHARACTERISTICS

30E D

Fig. 13 -- Output power vs. supply voltage

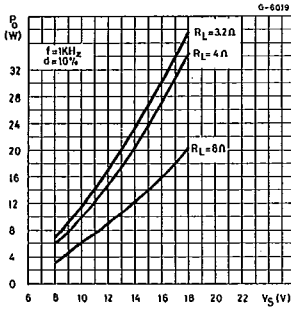


Fig. 14 - Distortion vs. frequency

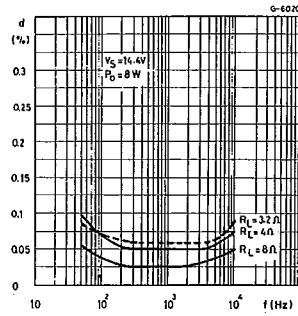


Fig. 15 - Supply voltage rejection vs. frequency

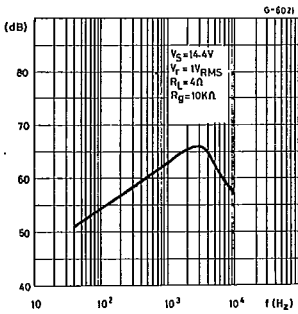


Fig. 16 - Power dissipation and efficiency vs. output power

